

10/12/99

## Solutions to sample problems for MT1. Pg. Venkaya

## 1. System

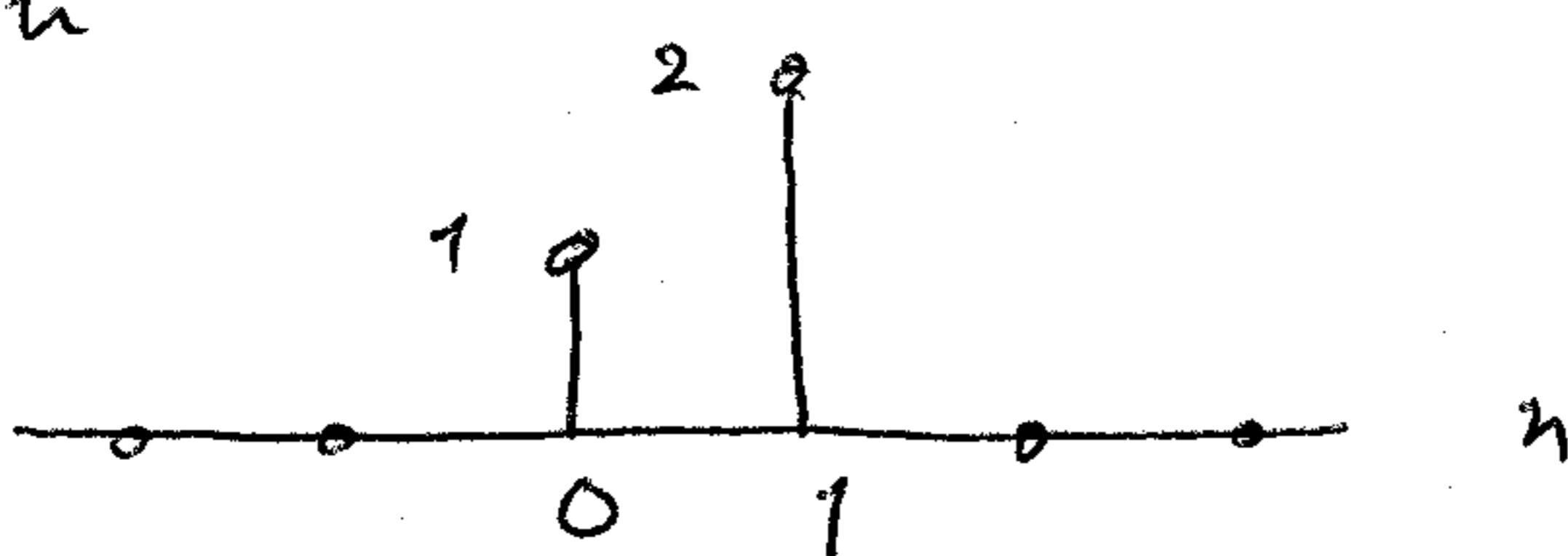
$$\begin{aligned} h_1(x)(t) &= x(t-1) \\ h_2(x)(t) &= x(-t) \\ h_3(x)(t) &= x(2t) \\ h_4(x)(t) &= x(t) + 1 \end{aligned}$$

	L	TI	Memoless	Causal
$h_1(x)(t)$	Y	Y	N	Y
$h_2(x)(t)$	Y	N	N	N
$h_3(x)(t)$	Y	N	N	N
$h_4(x)(t)$	N	Y	Y	Y

2.

$$h_n, \quad h(n) = \delta(n) + 2\delta(n-1)$$

so, 
$$h(n) = \begin{cases} 1, & n=0 \\ 2, & n=1 \\ 0, & \text{else} \end{cases} \quad (1)$$

a) Plot of  $h$ 

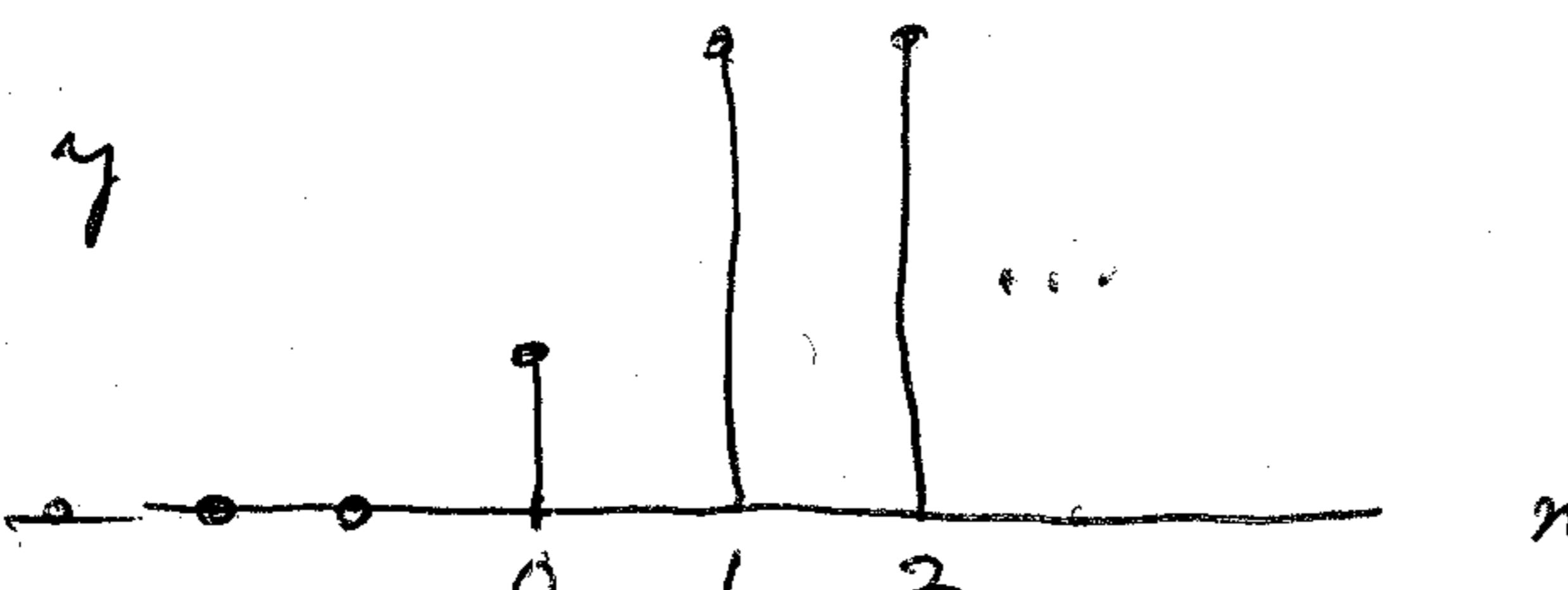
b) In general

$$y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= 1 x(n) + 2 x(n-1) \quad \text{from (1)}$$

$$= \begin{cases} 0, & n < 0 \\ 1, & n = 0 \\ 3, & n \geq 1 \end{cases}$$

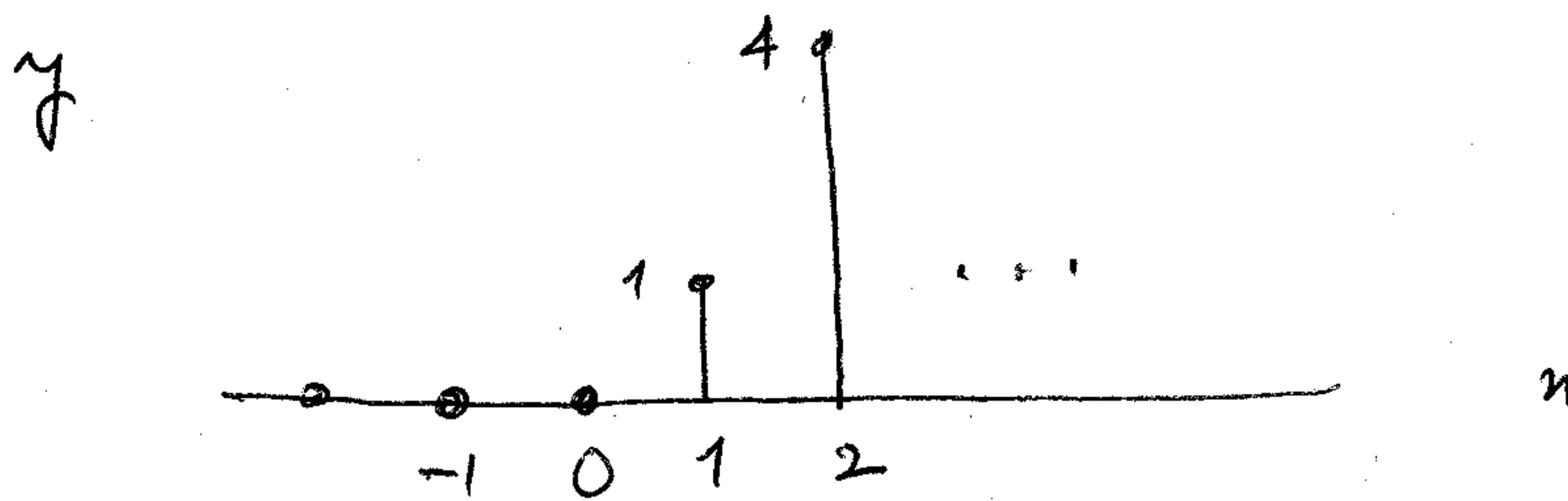
using the fact that  $x$  is unit step

Plot of  $y$ 

c) Using the form of  $x$

$$y(n) = 1 \cdot x(n) + 2 \cdot x(n-1)$$

$$= \begin{cases} 0 & , n < 0 \\ 0 & , n = 0 \\ n+2(n-1) = 3n-2 & , n \geq 1 \end{cases}$$



d) In general  $\hat{H}(\omega) = \text{DTFT}(h)$ ,

$$\begin{aligned} \hat{H}(\omega) &= \sum_{k=-\infty}^{\infty} h(k) e^{-ik\omega} \\ &= 1 + 2e^{-i\omega}, \quad \text{from (i)} \end{aligned}$$

e) For any  $\omega$

$$\begin{aligned} \hat{H}(\omega+2\pi) &= 1 + 2e^{-i(\omega+2\pi)} \\ &= 1 + 2e^{-i\omega} \cdot e^{-i2\pi} \\ &= 1 + 2e^{-i\omega}, \quad \text{since } e^{i2\pi} = 1 \\ &= \hat{H}(\omega) \end{aligned}$$

$$f) \hat{H}(\omega) = 1 + 2e^{-i\omega} = 1 + 2\cos\omega - 2i\sin\omega$$

$$\hat{H}(-\omega) = 1 + 2\cos(-\omega) - 2i\sin(-\omega)$$

$$= 1 + 2 \cos \omega + 2i \sin \omega$$

$$= [1 + 2 \cos \omega - 2i \sin \omega]^* = [H(\omega)]^*$$

g)  $\hat{H}(\omega) = [(1+2\cos\omega)^2 + 4\sin^2(\omega)]^{1/2}$

$$= [1 + 4 \cos \omega + 4 \cos^2 \omega + 4 \sin^2 \omega]^{1/2}$$

$$= [5 + 4 \cos \omega]^{1/2}$$

$$\cancel{\times} \hat{H}(\omega) = \tan^{-1} \left( -2 \sin \omega / (1 + 2 \cos \omega) \right)$$

$$= - \tan^{-1} \left( \frac{2 \sin \omega}{1 + 2 \cos \omega} \right)$$

h)  $|\hat{H}(-\omega)| = [5 + 4 \cos(-\omega)]^{1/2}$

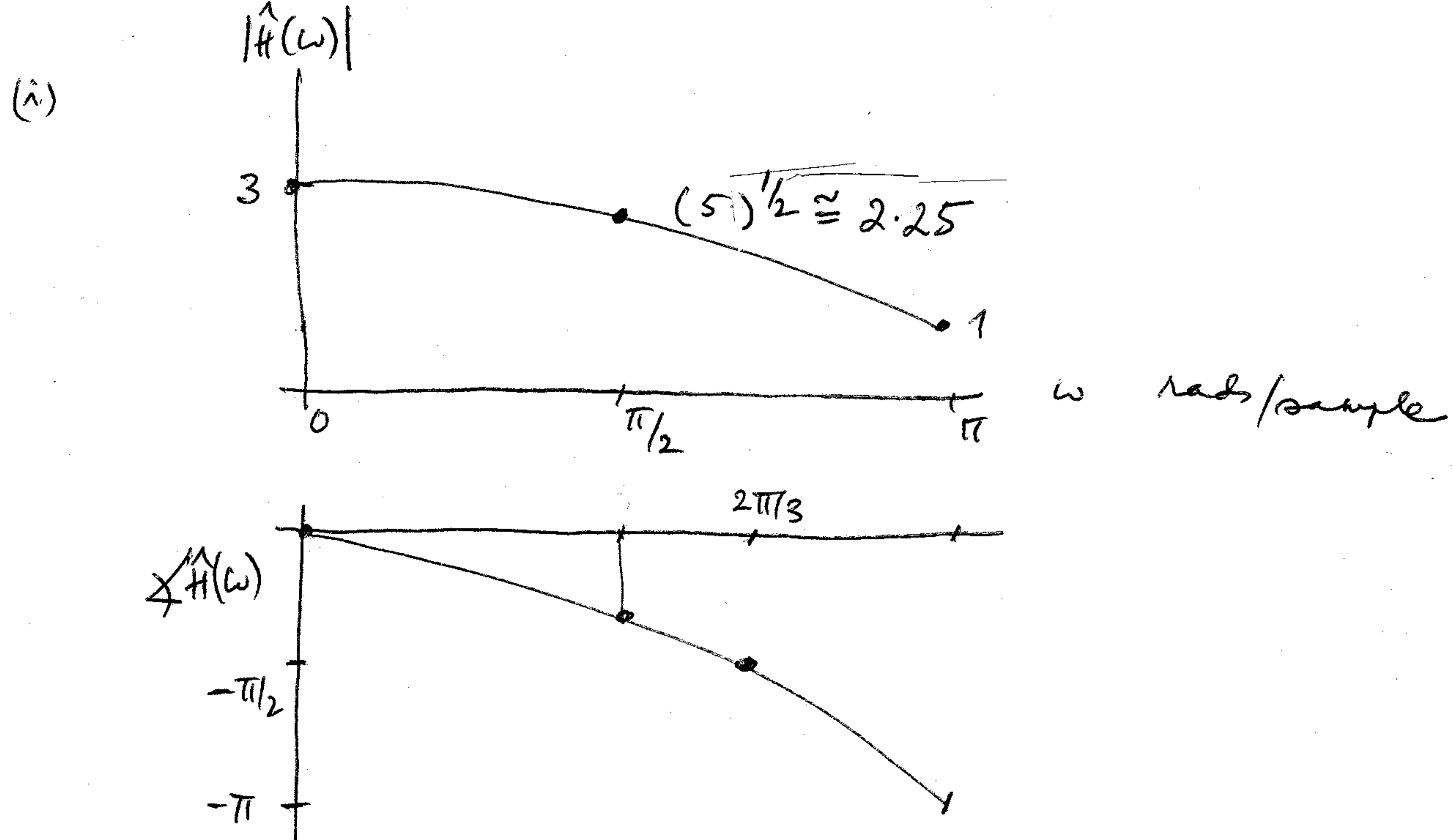
$$= [5 + 4 \cos \omega]^{1/2}, \text{ since } \cos(-\omega) = \cos \omega$$

$$= |\hat{H}(\omega)|$$

$$\cancel{\times} \hat{H}(-\omega) = - \tan^{-1} \left( \frac{2 \sin(-\omega)}{1 + 2 \cos(\omega)} \right)$$

$$= \tan^{-1} \left( \frac{2 \sin \omega}{1 + 2 \cos \omega} \right), \text{ since } \tan^{-1}(-x) = -\tan^{-1}x$$

$$= - \cancel{\times} \hat{H}(\omega)$$



To obtain these plots I use the following observations

$$(1) |\hat{H}(\omega)| = [5 + 4 \cos \omega]^{1/2}$$

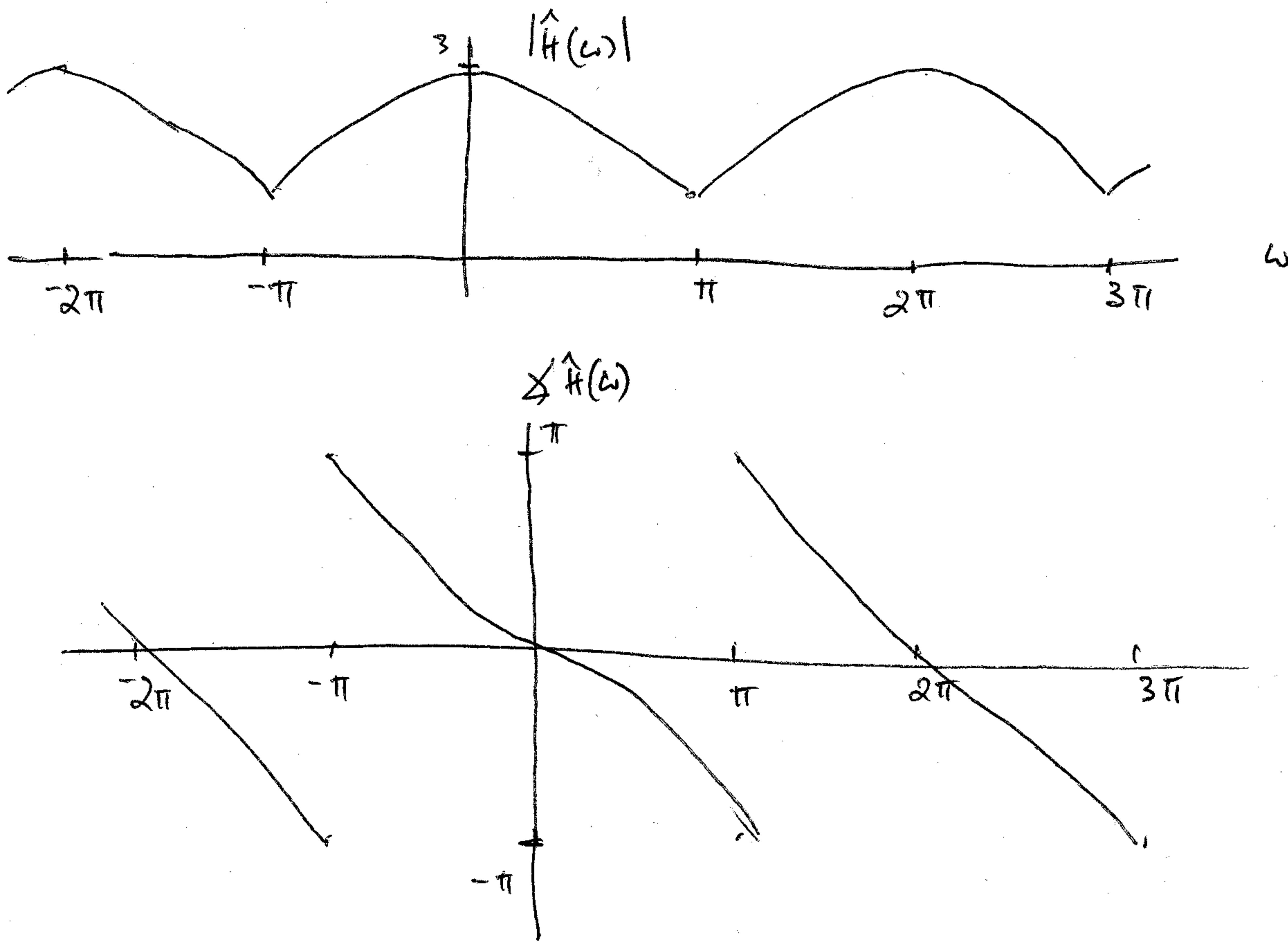
decreases as  $\omega$  goes from 0 to  $\pi$  because  $\cos \omega$  decreases. Then I use  $\cos 0 = 1$ ,  $\cos \frac{\pi}{4} = \frac{1}{\sqrt{2}}$ ,  $\cos \frac{\pi}{2} = 0$ ,  $\cos \pi = -1$  to get

$$\hat{H}(0) = 3, \quad \hat{H}\left(\frac{\pi}{4}\right) = (5+2\sqrt{2})^{1/2}, \quad \hat{H}\left(\frac{\pi}{2}\right) = 5^{1/2}, \quad \hat{H}(\pi) = 1$$

$$(2) \phi \hat{H}(\omega) = -\tan^{-1} \frac{2 \sin \omega}{1 + 2 \cos \omega}$$

$\sin \omega$  is positive for  $0 < \omega < \pi$  and  
 $(2 \cos \omega + 1) > 0$  for  $0 < \omega < \frac{2\pi}{3}$   
 $< 0$  for  $\frac{2\pi}{3} < \omega < \pi$

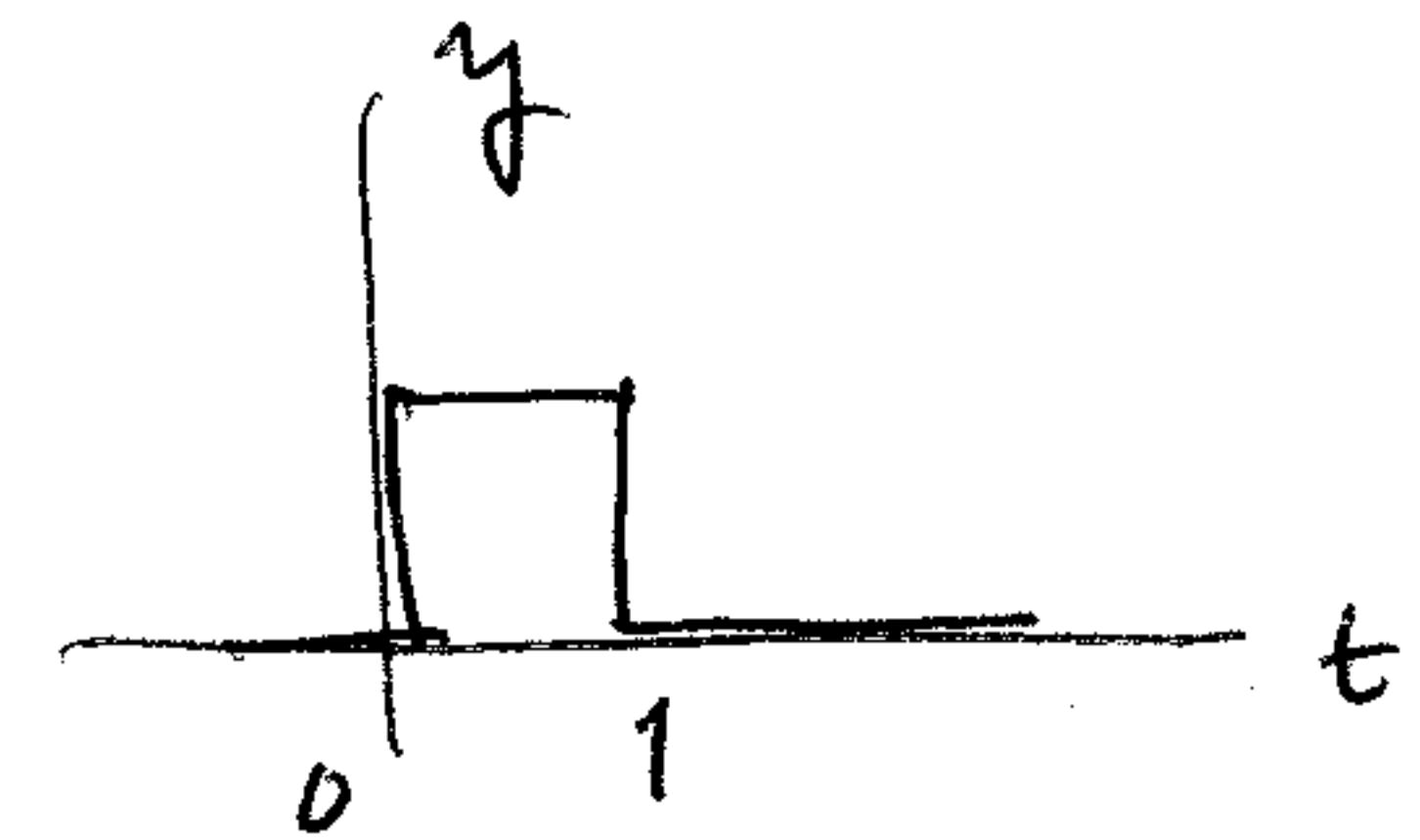
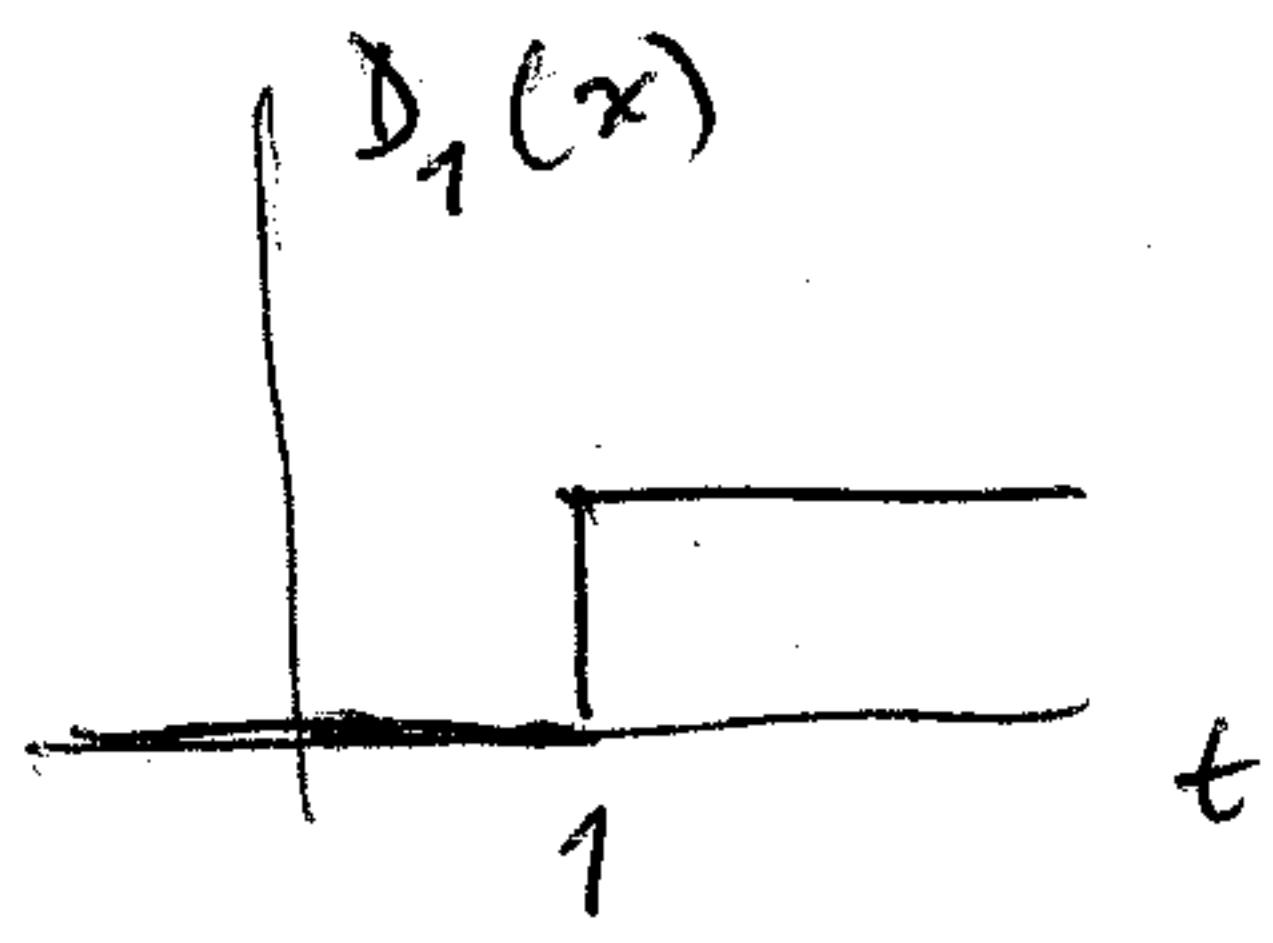
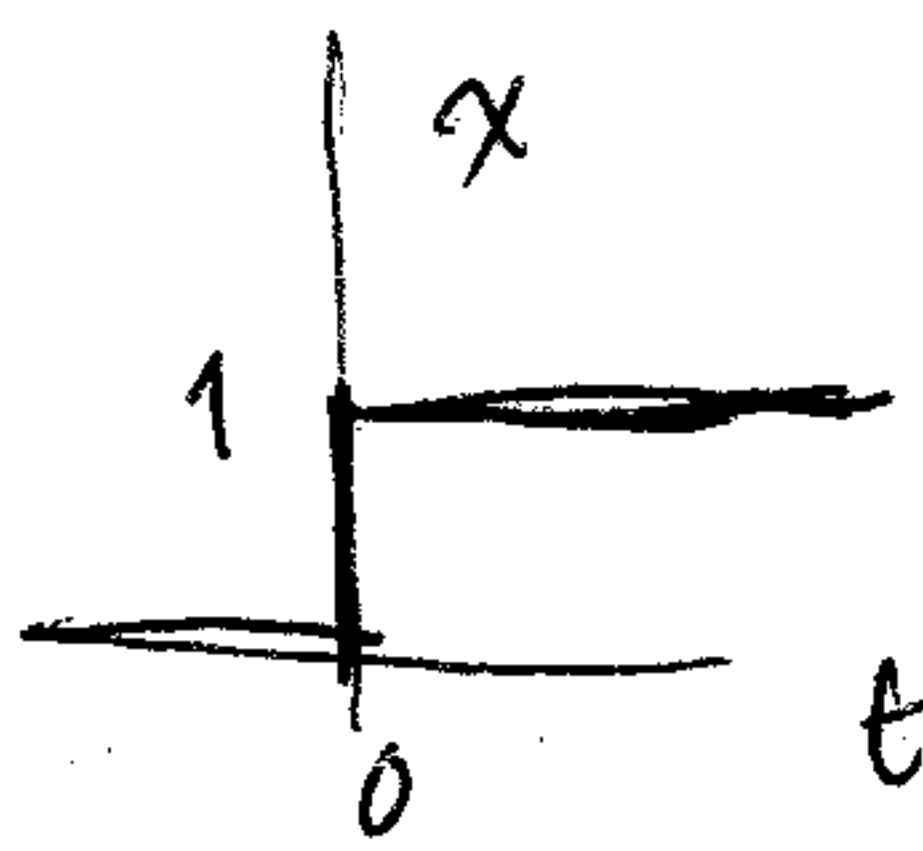
(j) Use the fact that  $\hat{H}(\omega)$  and  $\hat{x}(\omega)$  are periodic with period  $2\pi$ ,  $\hat{H}(\omega)$  is even function,  $\hat{x}(\omega)$  is odd to get following plots.



(k) In general, if the input is a sinusoidal signal of frequency  $\omega$ , the LTI system's response is an output sinusoid of same frequency, with magnitude multiplied by  $|\hat{H}(\omega)|$  and phase shifted by  $\hat{x}(\omega)$ . So in this case,

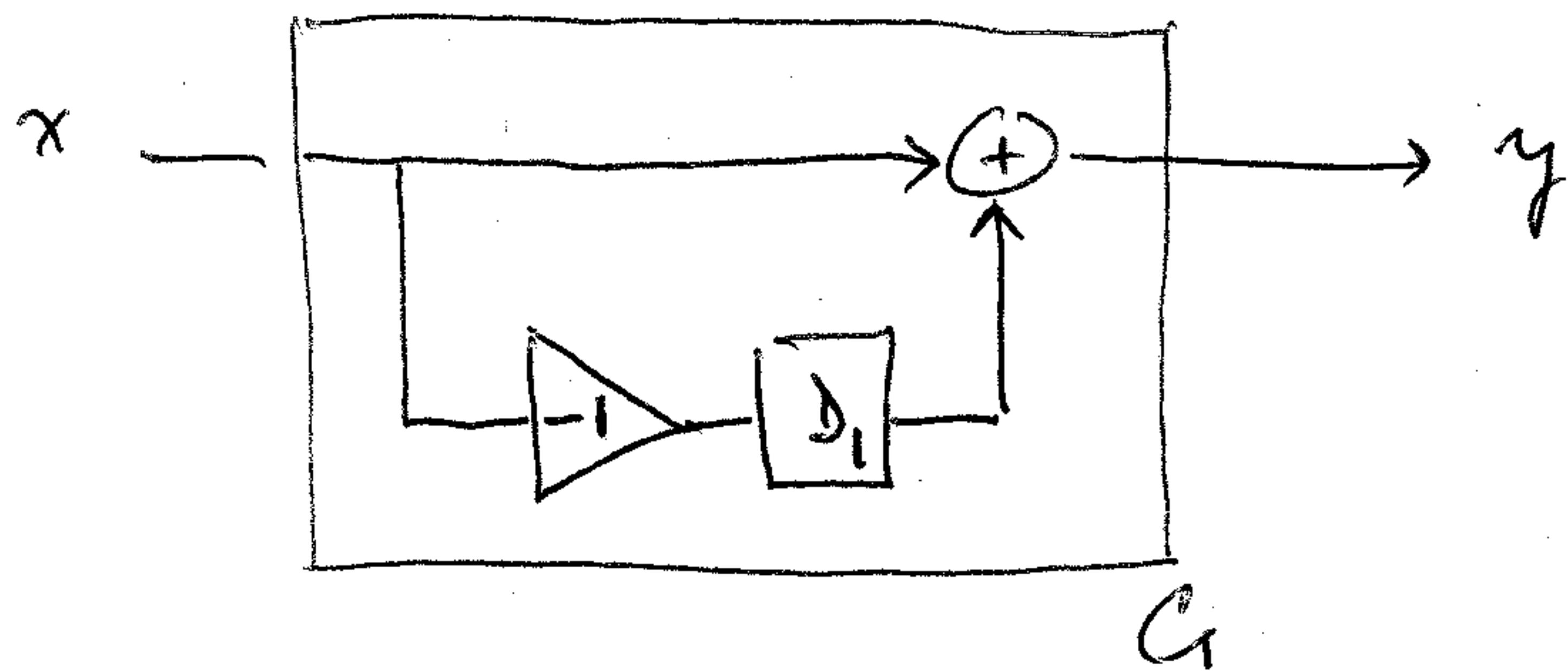
$$h_n, y(n) = |\hat{H}(25)| \cos \left( 25n + \frac{\pi}{6} + \hat{x}(25) \right) \\ + |\hat{H}(26)| \sin \left( 26n + \frac{\pi}{3} + \hat{x}(26) \right).$$

3.



$$a) \quad y = x - D_1(x) \quad (*)$$

b)

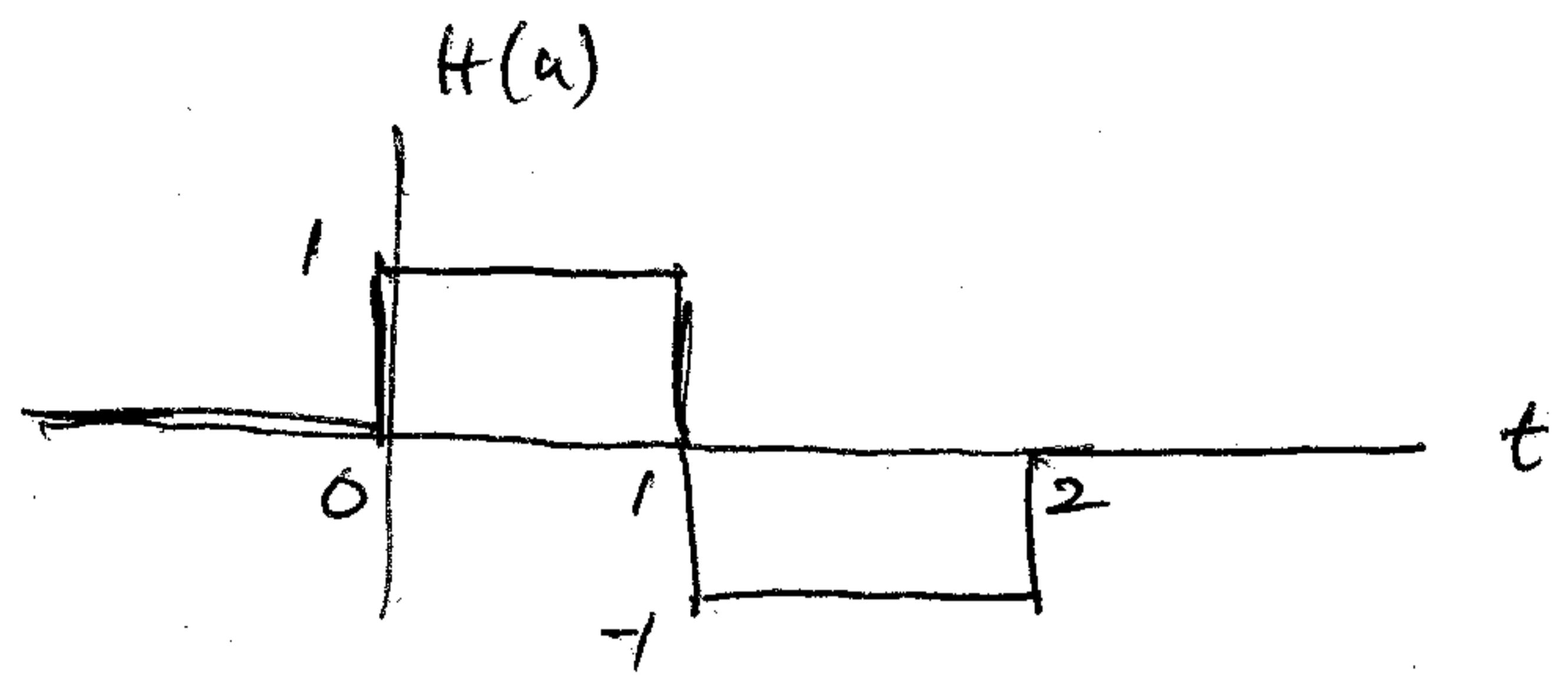


c) Call the input for the part  $u$ ,

$$u(t) = \begin{cases} 1, & 0 \leq t \leq 1 \\ 0, & \text{else.} \end{cases}$$

Then  $u = x - D_1(x)$  where  $x$  is the unit step. So the response to  $u$  is

$$\begin{aligned} H(u) &= H(x - D_1(x)) \\ &= H(x) - H(D_1(x)) \quad \text{by linearity} \\ &= H(x) - D_1 H(x) \quad \text{by time-invariance} \\ &= y - D_1(y) \\ &= (x - D_1(x)) - \overline{D_1(x - D_1(x))} \\ &= x - D_1(x) - D_1(x) + D_2(x) = x - 2D_1(x) + D_2(x) \end{aligned}$$



d) Since for all signals  $x$

$$G(x) = x - D_1(x)$$

we get

$$G(u) = u - D_1(u) = H(u)$$

e) To find  $\hat{G}(\omega)$ , take input

$$x(n) = e^{j\omega n}$$

The response of  $G$  to the input is

$$\begin{aligned} y(n) &= x(n) - D_1(x)(n) \\ &= e^{j\omega n} - e^{j\omega(n-1)} \\ &= [1 - e^{-j\omega}] e^{j\omega n} \\ &= \hat{G}(\omega) e^{j\omega n} \end{aligned}$$

So

$$\hat{G}(\omega) = 1 - e^{-j\omega}$$

4. Call the periodic function Saw. Then

$$\text{Saw}(t) = \sum_{-\infty}^{\infty} X_k e^{ik\omega t}$$

where  $\omega = \frac{2\pi}{T} = 2\pi$  ( $T=1$  sec, the period) and

$$X_k = \frac{1}{T} \int_0^T \text{Saw}(t) e^{-ikt} dt$$

$$= 2 \int_0^1 t e^{-ik2\pi t} dt \quad (*).$$

Now  $X_0 = 1$ , and for  $k \neq 0$

$$\int_0^1 t e^{-ik2\pi t} dt = \int_0^1 t d \left( \frac{e^{-ik2\pi t}}{-ik2\pi} \right)$$

$$= t \cdot \frac{e^{-ik2\pi t}}{-ik2\pi} \Big|_0^1 - \int_0^1 \frac{e^{-ik2\pi t}}{-ik2\pi} dt$$

$$= \frac{1}{-ik2\pi} - \frac{1}{(-ik2\pi)^2} e^{-ik2\pi t} \Big|_0^1$$

$$= \frac{i}{2\pi k} - \frac{1}{(-i2\pi k)^2} (1 - 1)$$

$$= \frac{i}{2\pi k}$$

Substituting in (\*) gives

$$S_{\text{aw}}(t) = 1 + \sum_{k=-\infty}^{\infty} \frac{2i}{2\pi k} e^{ik2\pi t} \quad k \neq 0$$

So the response of  $H$  to the signal  $S_{\text{aw}}$  is

$$\forall t, H(S_{\text{aw}})(t) = \hat{H}(0) \cdot 1 + \sum_{k \neq 0} \frac{2i}{2\pi k} \hat{H}(2\pi k) e^{i2\pi k t}$$

$$= \hat{H}(0) \cdot 1 = 1$$

since  $\hat{H}(2\pi k) = 0$  for  $k \neq 0$ .

5. a) The response  $y$  in general is

$$h_n y(n) = \sum_{k=-\infty}^{\infty} h(k) x(n-k)$$

$$= \sum_{k=0}^{\infty} \frac{1}{2^k} x(n-k)$$

$$= \begin{cases} \sum_{k=0}^n \frac{1}{2^k} = \frac{1 - (\frac{1}{2})^{n+1}}{1 - \frac{1}{2}}, & n \geq 0 \\ 0, & n < 0 \end{cases}$$

b) The frequency response is obtained by using the input

$$h_n x(n) = e^{j\omega n}$$

So

$$\hat{H}(\omega) e^{i\omega n} = \sum_{k=0}^{\infty} \frac{1}{2^k} e^{i\omega(n-k)}$$

$$= \left[ \sum_{k=0}^{\infty} \frac{1}{2^k} e^{-i\omega k} \right] e^{i\omega n}$$

So

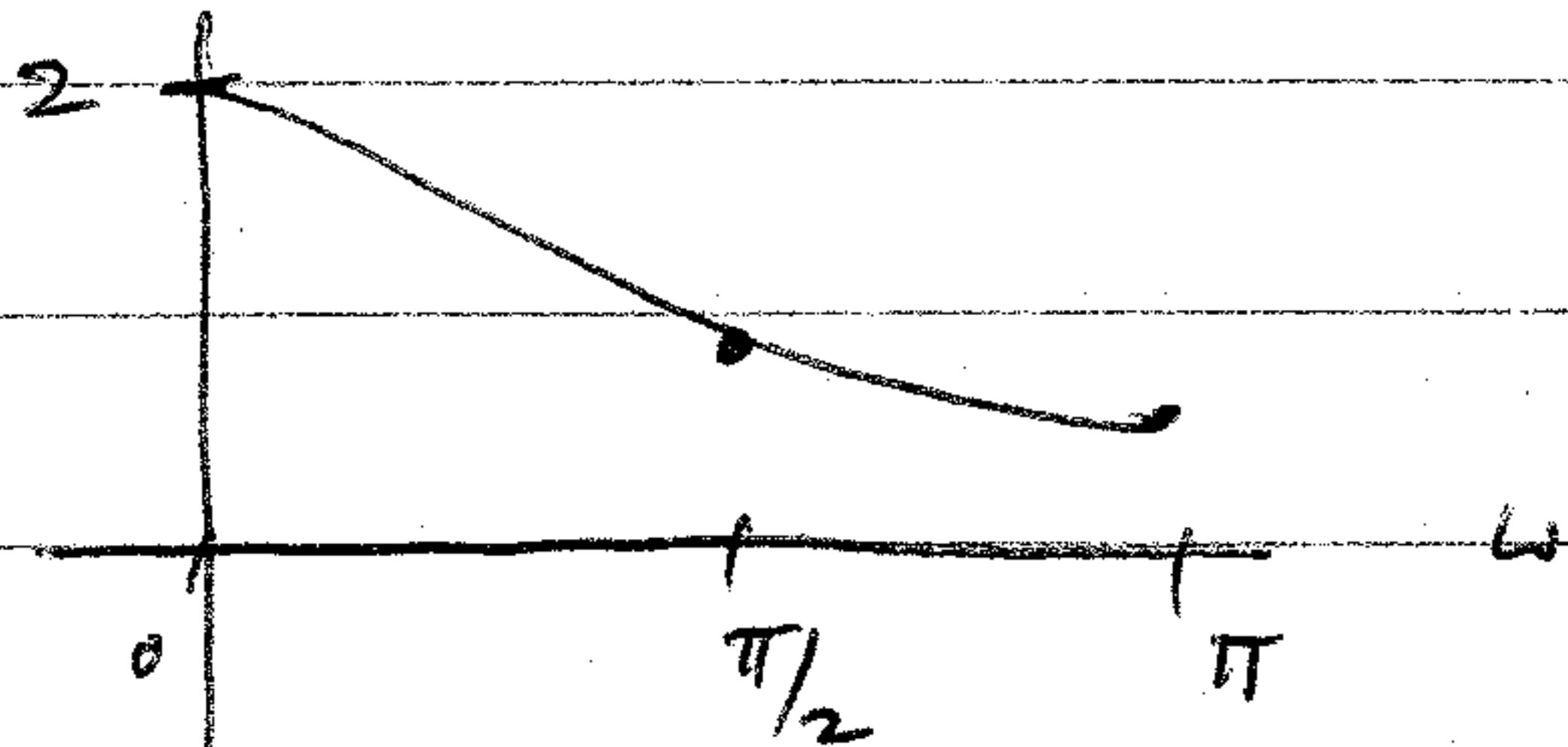
$$\hat{H}(\omega) = \sum_{k=0}^{\infty} \left( \frac{e^{-i\omega}}{2} \right)^k$$

$$= \frac{1}{1 - \frac{1}{2} e^{-i\omega}} = \frac{1}{1 - \frac{1}{2} \cos \omega + \frac{i}{2} \sin \omega}$$

$$|\hat{H}(\omega)| = \frac{1}{\left[ \left( 1 - \frac{1}{2} \cos \omega \right)^2 + \frac{1}{4} \sin^2 \omega \right]^{\frac{1}{2}}} \quad \times \hat{H}(\omega) = -\tan^{-1} \frac{\frac{1}{2} \sin \omega}{1 - \frac{1}{2} \cos \omega}$$

$$= \frac{1}{\left( \frac{5}{4} - \cos \omega \right)^{\frac{1}{2}}}$$

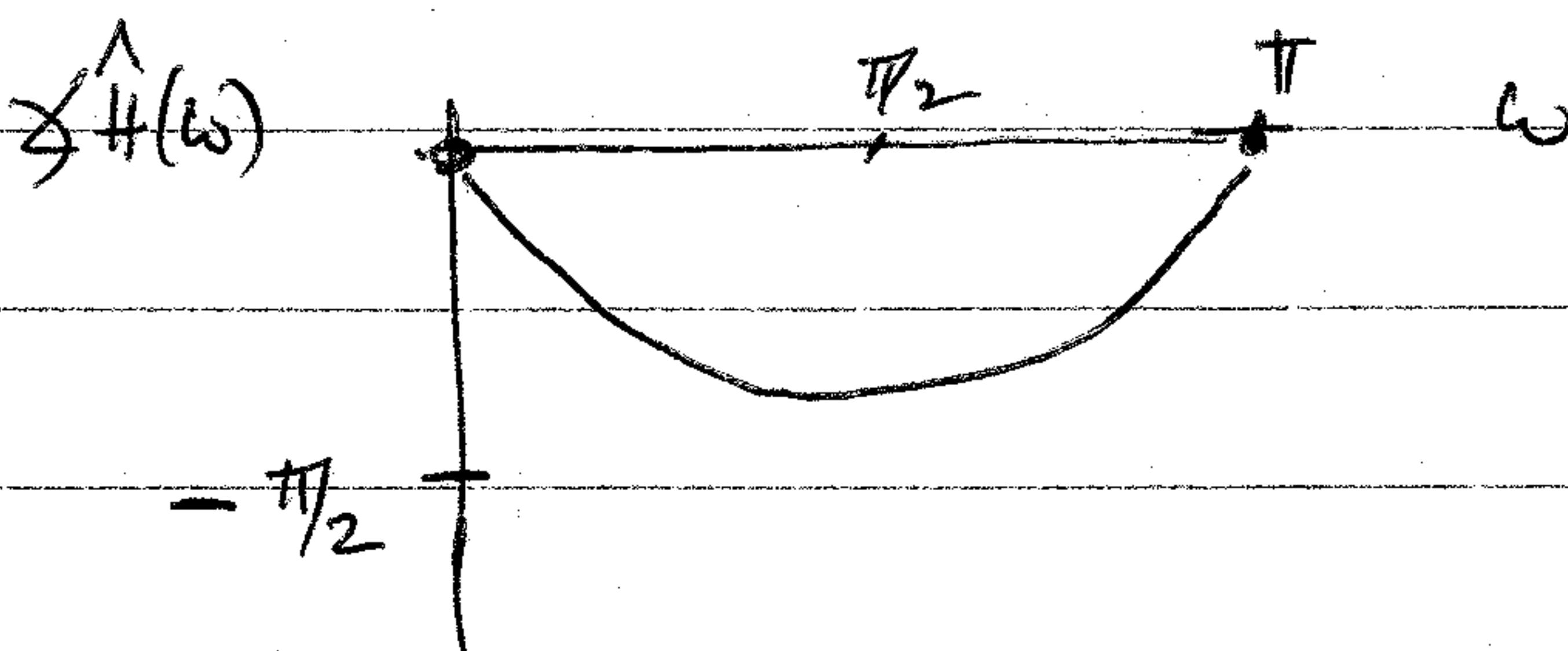
$\hat{H}(\omega)$



$$\text{Note } \hat{H}(0) = 2$$

$$\hat{H}(\pi/2) = 2/\sqrt{5}$$

$$\hat{H}(\pi) = 2/3$$



$$\text{Note } \hat{H}(0) = 0 = \hat{H}(\pi)$$

$$\hat{H}(\pi/2) = -\tan^{-1} \frac{1}{2}$$

c)  $\hat{G}(\omega) = [\hat{H}(\omega)]^2 = \left( \frac{1}{1 - 0.5e^{-i\omega}} \right)^2$

d) The closed-loop frequency response is

$$\frac{\hat{H}(\omega)}{1 + \hat{H}(\omega)} = \frac{1}{2 - 0.5e^{-i\omega}}$$